# Fundamentals of Mathematical Informatics 

Communication through Noisy Channels

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Lecture Four

## General communication scheme



## The discrete memoryless channel (DMC)

- In lecture one, we said that a RV $X$ is like a 'device' that outputs an element from a set $\left\{x_{1}, \cdots, x_{n}\right\}$ with probability $\operatorname{Pr}\left\{X=x_{i}\right\}=p_{i}$.
- Imagine now a 'device' that has an output and an input: it accepts strings of symbols from its input alphabet $\Sigma_{1}=\left\{a_{1}, \cdots, a_{m}\right\}$ and emits strings of symbols from an output alphabet $\Sigma_{2}=\left\{b_{1}, \cdots, b_{n}\right\}$.
- A discrete memoryless channel (DMC) is given by: an input alphabet $\Sigma_{1}=\left\{a_{1}, \cdots, a_{m}\right\}$, an output alphabet $\Sigma_{2}=\left\{b_{1}, \cdots, b_{n}\right\}$, and a channel matrix $P=\llbracket p_{i j} \rrbracket_{i j}(1 \leqslant i \leqslant m, 1 \leqslant j \leqslant n)$ of transition probabilities:

$$
p_{i j} \stackrel{\text { def }}{=} p\left(b_{j} \mid a_{i}\right) \stackrel{\text { def }}{=} \operatorname{Pr}\left\{\text { output is } b_{j} \mid \text { input was } a_{i}\right\} .
$$

Therefore, $p_{i j} \geqslant 0$ for all $i$ and $j$, and $\sum_{j} p_{i j}=1$ for all $i$.

- Memory trick. To remember which is the input and which is the output, think as if $p_{i j}=p_{i \rightarrow j}$.
- The channel is 'discrete' because input and output alphabets are discrete sets.
- The channel is 'memoryless' because the channel matrix $P$ remains the same for repeated uses.


## Example: the binary erasure channel



In this case, $\Sigma_{1}=\{0,1\}, \Sigma_{2}=\{0,1, \otimes\}$, and

$P=$|  | 0 | 1 | $\odot$ |
| :---: | :---: | :---: | :---: |
| 0 | $1-\epsilon$ | 0 | $\epsilon$ |
| 1 | 0 | $1-\epsilon$ | $\epsilon$ |

## Example: the binary symmetric channel



In this case, $\Sigma_{1}=\Sigma_{2}=\{0,1\}$ and

$P=$|  | 0 | 1 |
| :---: | :---: | :---: |
| 0 | $1-p$ | $p$ |
| 1 | $p$ | $1-p$ |

## Input and output of a DMC as RVs

- Let $X$ be a RV with range $\mathcal{X}=\left\{x_{1}, \cdots, x_{m}\right\}$ and probability distribution $\left(p_{1}, \cdots, p_{m}\right)$.
- Take now a $\operatorname{DMC} \mathcal{N}$ with input alphabet $\mathcal{X}$, output alphabet $\mathcal{Y}=\left\{y_{1}, \cdots, y_{n}\right\}$, and channel matrix $P=\llbracket p_{i j} \rrbracket$.
- What happens if we 'feed' $X$ through $\mathcal{N}$ ?
- $\operatorname{Pr}\left\{{ }^{\prime}\right.$ output is $\left.y_{j}{ }^{\prime}\right\}=\sum_{i=1}^{m} \operatorname{Pr}\left\{X=x_{i}\right\} p_{i j}=\sum_{i=1}^{m} p_{i} p_{i j}$.
- We obtain another RV $Y$, with range equal to $\mathcal{Y}$ and probability distribution $\left(q_{1}, \cdots, q_{n}\right)$ where $q_{j}=\sum_{i} p_{i} p_{i j}$.


## I/O joint distribution

With the notation introduced above, the action of a DMC channel $\mathcal{N}$ on an input $\mathrm{RV} X$ gives rise to a pair of dependent $\mathrm{RVs}(X, Y)$ with joint probability distribution given by

$$
\operatorname{Pr}\left\{X=x_{i} \text { and } Y=y_{j}\right\}=p_{i} p_{i j}
$$

Sometimes we write $Y=\mathcal{N}(X)$.

## $r$-th extension of a DMC: in series

What happens when we feed a string of $r$ symbols $\left(\alpha_{1}, \cdots, \alpha_{r}\right) \in \Sigma_{1}^{(r)}$ through a discrete memoryless channel?


The $r$-th extension of a DMC is then itself a DMC from an input $r$-dimensional RV $\boldsymbol{X}$ with range $\Sigma_{1}^{(r)}$, to an output $r$-dimensional RV $\boldsymbol{Y}$ with range $\Sigma_{2}^{(r)}$. The channel matrix is given by the product of the transition probabilities:
$\operatorname{Pr}\left\{\boldsymbol{Y}=\beta_{1} \cdots \beta_{r} \mid \boldsymbol{X}=\alpha_{1} \cdots \alpha_{r}\right\} \stackrel{\text { def }}{=} p(\boldsymbol{Y} \mid \boldsymbol{X})=p\left(\beta_{1} \mid \alpha_{1}\right) \cdots p\left(\beta_{r} \mid \alpha_{r}\right)$.

## $r$-th extension of a DMC: in parallel

We can also think of channel extensions this way:


We have now many copies of the same noisy channel acting 'in parallel.' Mathematically, serial and parallel extensions are equivalent.

## Example: sending a message through a binary symmetric channel

- Imagine that we want to send one word $s_{i}$, chosen at random among eight possible words $\left\{s_{1}, \cdots, s_{8}\right\}$, via a binary symmetric DMC.
- First, we have to encode all words in binary alphabet (the channel only accepts 0 s and $1 \mathrm{~s}!$ ).
- $s_{1} \mapsto 000, s_{2} \mapsto 001, \cdots, s_{8} \mapsto 111$.
- Here we use the third extension of the binary symmetric channel (the input consists of three bits.)
- What is the probability that the receiver gets the wrong word?
$\operatorname{Pr}\{$ wrong word $\}=1-\operatorname{Pr}\{$ correct word $\}=1-(1-p)^{3}=$
$p\left(3-3 p+p^{2}\right)$. (For $p=0.5$ is $\approx 0.88$; for $p=0.1$ is $\approx 0.27$.)
- Can we do better?


## First idea: repetition codes (repeating words)

- Let's try to send each word twice through the channel, i.e.,

$$
s_{1} \mapsto 000000, s_{2} \mapsto 001001, \cdots, s_{8} \mapsto 111111
$$

- As a decoding rule, if the receiver does not get the same word twice in succession, she requests an immediate resending.
- What is the probability of decoding error in this case, i.e., the probability that the receiver gets the wrong word without detecting it?
- First possibility: one error in the first three bits and one error, in the same position, in the second three bits. This contributes with $3 \times p(1-p)^{2} \times p(1-p)^{2}=3 p^{2}(1-p)^{4}$.
- Second possibility: two errors in the first three bits, and two errors, in the same positions, in the second three bits. This contributes with $3 \times p^{2}(1-p) \times p^{2}(1-p)=3 p^{4}(1-p)^{2}$.
- Third possibility: six errors in a row. This contributes with $p^{6}$.
- Total decoding error probability: $p^{2}\left(3-12 p+21 p^{2}-18 p^{3}+7 p^{4}\right)$. (For $p=0.5$ is $\approx 0.11$; for $p=0.1$ is $\approx 0.02$.)
- But: it requires feedback from the receiver, for each letter sent.
- But: with increasing length, the receiver will almost always request a resending.
- Hence: zero total decoding error requires infinite repetitions (no reliable communication is possible)


## Second idea: parity-check codes

- Instead of just repeating codewords, we can try to exploit another idea.
- Parity-check coding: it adds one extra bit (the 'parity bit') at the end of each codeword, so that the sum of the digits is always even.
- In our case, this gives:
$s_{1} \mapsto 0000, s_{2} \mapsto 0011, s_{3} \mapsto 0101, s_{4} \mapsto 0110, \cdots s_{8} \mapsto 1111$.
- If the receiver gets four bits whose sum is odd, she requests an immediate resending. (Hence the name, 'parity-check.')
- What is the probability of decoding error in this case, i.e., the probability that the receiver gets the wrong word without detecting it?
- A wrong decoding happens if there were two or four errors, therefore the decoding error probability is $6 p^{2}(1-p)^{2}+p^{4}$. (For $p=0.5$ is $\approx 0.44$; for $p=0.1$ is $\approx 0.05$.)
- But: this code requires feedback.
- Remark: this simple idea can be improved, and it is at the basis of some very important families of codes (Low Density Parity-Check, LDPC).


## Third idea: Shannon approach (definitions)

- Take a DMC with $\Sigma_{1}$ and $\Sigma_{2}$ as input and output alphabets, respectively.
- An $(M, n)$ code consists of the following:
(1) An index set $\{1,2, \cdots, M\}$.
(2) An encoding function $\boldsymbol{c}:\{1,2, \cdots, M\} \rightarrow \Sigma_{1}^{(n)}$ (i.e., each $\boldsymbol{c}_{i} \stackrel{\text { def }}{=} \boldsymbol{c}(i)$ is a string of $n$ symbols in $\Sigma_{1}$, e.g., $\left.\boldsymbol{c}_{i}=\alpha_{1} \alpha_{2} \cdots \alpha_{n}\right)$.
(3) A decoding function $g: \Sigma_{2}^{(n)} \rightarrow\{1,2, \cdots, M\}$.
- The collection $\mathscr{C}=\left\{\boldsymbol{c}_{1}, \cdots, \boldsymbol{c}_{M}\right\}$ is called the codebook and its elements are called the codewords. $M$ (the number of codewords) is the size of the code, while $n$ (the length of each codeword) is its length.


## Encoding-transmission-decoding: chain of RVs

The encoding-transmission-decoding process can be summarized as:

$$
W \xrightarrow{\mathcal{E}_{n}} \boldsymbol{X} \xrightarrow{\mathcal{N}_{n}} \boldsymbol{Y} \xrightarrow{\mathcal{D}_{n}} \hat{W}
$$

What does this mean?

- $W$, the message: a RV with range $\left\{w_{1}, \cdots, w_{M}\right\}$ and probabilities $p_{1}, \cdots p_{M}$.
- $\mathcal{E}_{n}: W \rightarrow \boldsymbol{X}$, the length- $n$ encoding: a DMC with input alphabet $\left\{w_{1}, \cdots, w_{M}\right\}$, output alphabet $\Sigma_{1}^{(n)}$, and channel matrix given by $p\left(\boldsymbol{X} \mid w_{i}\right) \stackrel{\text { def }}{=} \operatorname{Pr}\left\{\boldsymbol{X}=\alpha_{1} \cdots \alpha_{n} \mid W=w_{i}\right\}=\delta_{\boldsymbol{X}, c_{i}}$.
- $\mathcal{N}_{n}: \boldsymbol{X} \rightarrow \boldsymbol{Y}$, the $n$-th extension of the communication channel: a DMC with input alphabet $\Sigma_{1}^{(n)}$, output alphabet $\Sigma_{2}^{(n)}$, and channel matrix $p(\boldsymbol{Y} \mid \boldsymbol{X}) \stackrel{\text { def }}{=} \operatorname{Pr}\left\{\boldsymbol{Y}=\beta_{1} \cdots \beta_{n} \mid \boldsymbol{X}=\alpha_{1} \cdots \alpha_{n}\right\}=p\left(\beta_{1} \mid \alpha_{1}\right) \cdots p\left(\beta_{n} \mid \alpha_{n}\right)$.
- $\mathcal{D}_{n}: \boldsymbol{X} \rightarrow \hat{W}$, the decoding: a DMC with input alphabet $\Sigma_{2}^{(n)}$, output alphabet $\left\{w_{1}, \cdots, w_{M}\right\}$, and channel matrix given by $p\left(w_{j} \mid \boldsymbol{Y}\right) \stackrel{\text { def }}{=} \operatorname{Pr}\left\{\hat{W}=w_{j} \mid \boldsymbol{Y}=\beta_{1} \cdots \beta_{n}\right\}=\delta_{g\left(\beta_{1} \cdots \beta_{n}\right), j}$.
- A decoding error happens whenever $\hat{W} \neq W$. What is the probability that a decoding error occurs?


## A picture



## Decoding error probability

- How to compute the error probability, i.e. $\operatorname{Pr}\{\hat{W} \neq W\}$ ?

$$
\begin{aligned}
\operatorname{Pr}\{\hat{W} \neq W\} & \stackrel{\text { def }}{=} \sum_{j \neq i} \sum_{i=1}^{M} \operatorname{Pr}\left\{\hat{W}=w_{j}, W=w_{i}\right\} \\
& =\sum_{i, j=1}^{M} \operatorname{Pr}\left\{\hat{W}=w_{j}, W=w_{i}\right\}-\sum_{i=1}^{M} \operatorname{Pr}\left\{\hat{W}=w_{i}, W=w_{i}\right\} \\
& =1-\sum_{i=1}^{M} \operatorname{Pr}\left\{\hat{W}=w_{i}, W=w_{i}\right\} \\
& =1-\sum_{i} \sum_{\boldsymbol{X}} \sum_{\boldsymbol{Y}} p\left(w_{i} \mid \boldsymbol{Y}\right) p(\boldsymbol{Y} \mid \boldsymbol{X}) p\left(\boldsymbol{X} \mid w_{i}\right) p_{i} \\
& =1-\sum_{i} \sum_{\boldsymbol{X}} \sum_{\boldsymbol{Y}} \delta_{g(\boldsymbol{Y}), i} p(\boldsymbol{Y} \mid \boldsymbol{X}) \delta_{\boldsymbol{X}, \boldsymbol{c}_{i}} p_{i} \\
& =1-\sum_{i} \sum_{\boldsymbol{Y} \in g^{-1}(i)} p\left(\boldsymbol{Y} \mid \boldsymbol{c}_{i}\right) p_{i}
\end{aligned}
$$

- The error probability crucially depends on the choice of the decoding function $g$.


## Ideal-observer (minimum error) decoding

- $\operatorname{Pr}\{\hat{W} \neq W\}=1-\sum_{j} \sum_{\boldsymbol{Y}} \delta_{j, g(\boldsymbol{Y})} p\left(\boldsymbol{Y} \mid \boldsymbol{c}_{j}\right) p_{j}$.
- Rewrite $p\left(\boldsymbol{Y} \mid \boldsymbol{c}_{j}\right) p_{j}$ as $p\left(\boldsymbol{c}_{j}, \boldsymbol{Y}\right) \stackrel{\text { def }}{=} \operatorname{Pr}\left\{\boldsymbol{c}_{j}\right.$ sent and $\boldsymbol{Y}$ received $\}$.
- Rewrite it again as $p\left(\boldsymbol{c}_{j}, \boldsymbol{Y}\right)=p\left(\boldsymbol{c}_{j} \mid \boldsymbol{Y}\right) p_{\boldsymbol{Y}}$, where $p_{\boldsymbol{Y}} \stackrel{\text { def }}{=} \operatorname{Pr}\{\boldsymbol{Y}$ received $\}=\sum_{j=1}^{M} p\left(\boldsymbol{c}_{j}, \boldsymbol{Y}\right)$.
- Then, $\operatorname{Pr}\{\hat{W} \neq W\}=1-\sum_{\boldsymbol{Y}} \sum_{j} \delta_{j, g(\boldsymbol{Y})} p\left(\boldsymbol{c}_{j} \mid \boldsymbol{Y}\right) p_{\boldsymbol{Y}}$.
- Choose the decoding function $g: \Sigma_{2}^{(n)} \rightarrow\{1,2, \cdots, M\}$ in such a way that $p\left(\boldsymbol{c}_{g(\boldsymbol{Y})} \mid \boldsymbol{Y}\right) \geqslant p\left(\boldsymbol{c}_{j} \mid \boldsymbol{Y}\right)$, for all $1 \leqslant j \leqslant M$.
- Equivalently: $g(\boldsymbol{Y})=\arg \max _{j} p\left(\boldsymbol{c}_{j} \mid \boldsymbol{Y}\right)$.
- This decoding method is called ideal-observer or minimum-error, because it minimizes the error probability.
- Meaning: upon receiving $\boldsymbol{Y}$, use this piece of information to infer the most probable codeword.
- The ideal-observer decoding is optimal! However: the construction depends on the choice of probabilities $p_{1}, \cdots, p_{M}$, which is a serious disadvantage.


## Maximum-likelihood decoding

- $\operatorname{Pr}\{\hat{W} \neq W\}=1-\sum_{j} \sum_{\boldsymbol{Y}} \delta_{j, g(\boldsymbol{Y})} p\left(\boldsymbol{Y} \mid \boldsymbol{c}_{j}\right) p_{j}$.
- Choose a decoding function $g: \Sigma_{2}^{(n)} \rightarrow\{1,2, \cdots, M\}$ such that $p\left(\boldsymbol{Y} \mid \boldsymbol{c}_{g(\boldsymbol{Y})}\right) \geqslant p\left(\boldsymbol{Y} \mid \boldsymbol{c}_{j}\right)$, for all $1 \leqslant j \leqslant M$.
- Equivalently: $g(\boldsymbol{Y})=\arg \max _{j} p\left(\boldsymbol{Y} \mid \boldsymbol{c}_{j}\right)$.
- This decoding method is called maximum-likelihood (ML).
- Meaning: upon receiving $\boldsymbol{Y}$, decode it with the codeword $\boldsymbol{c}_{i}$ that, if sent, maximizes the probability of receiving $\boldsymbol{Y}$.
- Since, in general, $p\left(\boldsymbol{Y} \mid \boldsymbol{c}_{i}\right) \neq p\left(\boldsymbol{c}_{i} \mid \boldsymbol{Y}\right)$, ML decoding and ideal-observer decoding may give different results.
- Con: sub-optimal. Pro: independent of the $p_{i}$ 's, much easier to implement.
- Question: when do ML and ideal-observer decodings agree?

Answer: they agree if $p_{1}=p_{2}=\cdots=p_{M}=\frac{1}{M}$.

## Example: minimum-error vs max-likelihood

- Suppose you are at the receiver's end of a binary symmetric channel with error probability $\epsilon \stackrel{\text { def }}{=} p_{0 \rightarrow 1}=p_{1 \rightarrow 0}=\frac{9}{10}$.
- Suppose you receive a 'zero.' What is the best guess for the input?
- Since the channel introduce an error $90 \%$ of the times, one would say: the best guess is that the input was 'one.'
- This is what a max-likelihood strategy says.
- However, imagine that you know that the sender sends 'zero' with probability $p(\mathrm{in}=0)=\frac{19}{20}$ and 'one' with $p(\mathrm{in}=1)=\frac{1}{20}$.
- Then, $p($ in $=1 \mid$ out $=0)=\frac{p(\mathrm{in}=1 \text { and out }=0)}{p(\text { out }=0)}=\frac{\epsilon p(\mathrm{in}=1)}{(1-\epsilon) p(\mathrm{in}=0)+\epsilon p(\mathrm{in}=1)}=$ $\frac{\frac{9}{10} \frac{1}{20}}{\frac{1}{10} \frac{19}{20}+\frac{9}{10} \frac{1}{20}}=\frac{9}{28} \approx 0.32$.
- Therefore $p(\mathrm{in}=0 \mid$ out $=0)=1-p(\mathrm{in}=1 \mid$ out $=0) \approx 0.68$.
- According to the ideal-observer rule (the optimal one), the best guess is that the input was 'zero.'


## Minimum-distance decoding (Hamming distance)

- Let $\mathrm{V}_{n}$ be the set of all binary sequences of length $n$.
- Definition: given $\boldsymbol{x}, \boldsymbol{y} \in \mathrm{V}_{n}$, their Hamming distance $d(\boldsymbol{x}, \boldsymbol{y})$ is defined as the number of places in which $\boldsymbol{x}$ and $\boldsymbol{y}$ differ.
- Example: take $\mathrm{V}_{4}$ and $\boldsymbol{x}=0001$ and $\boldsymbol{y}=1011$; then $d(\boldsymbol{x}, \boldsymbol{y})=2$ (first and third digits are different).
- Minimum-distance decoding: choose the decoding function $g: \Sigma_{2}^{(n)} \rightarrow\{1, \cdots, M\}$ such that $d\left(\boldsymbol{Y}, \boldsymbol{c}_{g(\boldsymbol{Y})}\right) \leqslant d\left(\boldsymbol{Y}, \boldsymbol{c}_{j}\right)$, for all $1 \leqslant j \leqslant M$.
- Meaning: upon receiving $\boldsymbol{Y}$, decode it with a codeword $\boldsymbol{c}_{i}$ that is 'as close as possible' to $\boldsymbol{Y}$, according to the Hamming distance.


## Min-Distance $\equiv$ Max-Likelihood (for binary symmetric channels)

Proof. Let $\epsilon \leqslant 1 / 2$ the bit-flip probability of the channel. For any $\boldsymbol{x}, \boldsymbol{y} \in \mathrm{V}_{n}$ with $d(\boldsymbol{x}, \boldsymbol{y})=k$,

$$
\operatorname{Pr}\{\boldsymbol{y} \text { received } \mid \boldsymbol{x} \text { sent }\}=\epsilon^{k}(1-\epsilon)^{n-k},
$$

which is maximum when $k$ is minimum.


## Summary of lecture four

- Discrete memoryless channels provide a simple (but very important) model of communication channels
- The coding problem is to design encoding-decoding methods that allow the receiver to guess (with high reliability) the correct input, avoiding errors.
- The optimal decoding method is called ideal-observer decoding, but it is not practical.
- The maximum-likelihood and the minimum-distance decoding are preferable.


## Keywords of lecture four

discrete memoryless channel, binary symmetric channel, $r$-th extension of a DMC, repetition codes, parity-check codes, encoding-transmission-decoding scheme, decoding error probability, ideal-observer decoding, maximum-likelihood decoding, Hamming distance, minimum-distance decoding

